The background of the slide is a repeating pattern of green leaf outlines on a white background. The leaves are stylized and arranged in a way that creates a sense of depth and movement.

Engineering  
and  
Architecture  
Analog  
Electronics  
Fundamentals  
**EEE223**

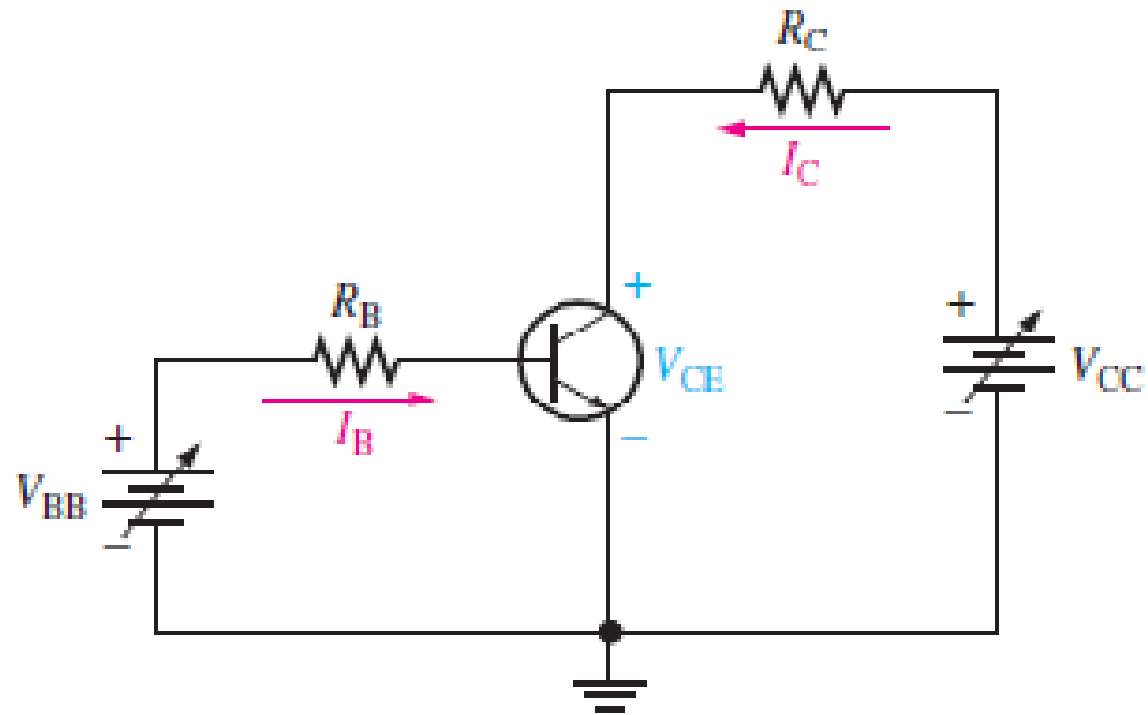
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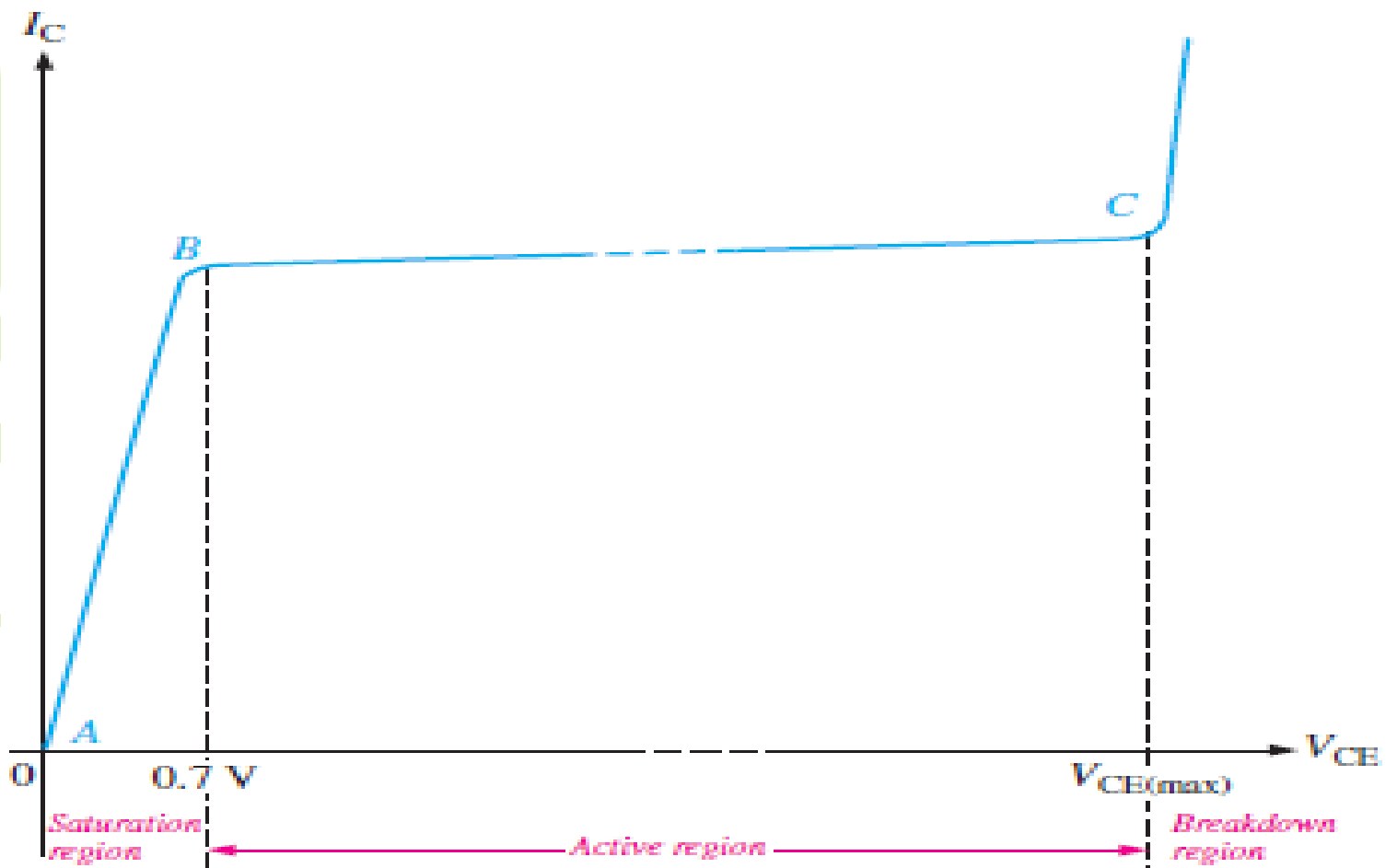
Bipolar Junction Transistor BJT

Lecturer Sally Adil

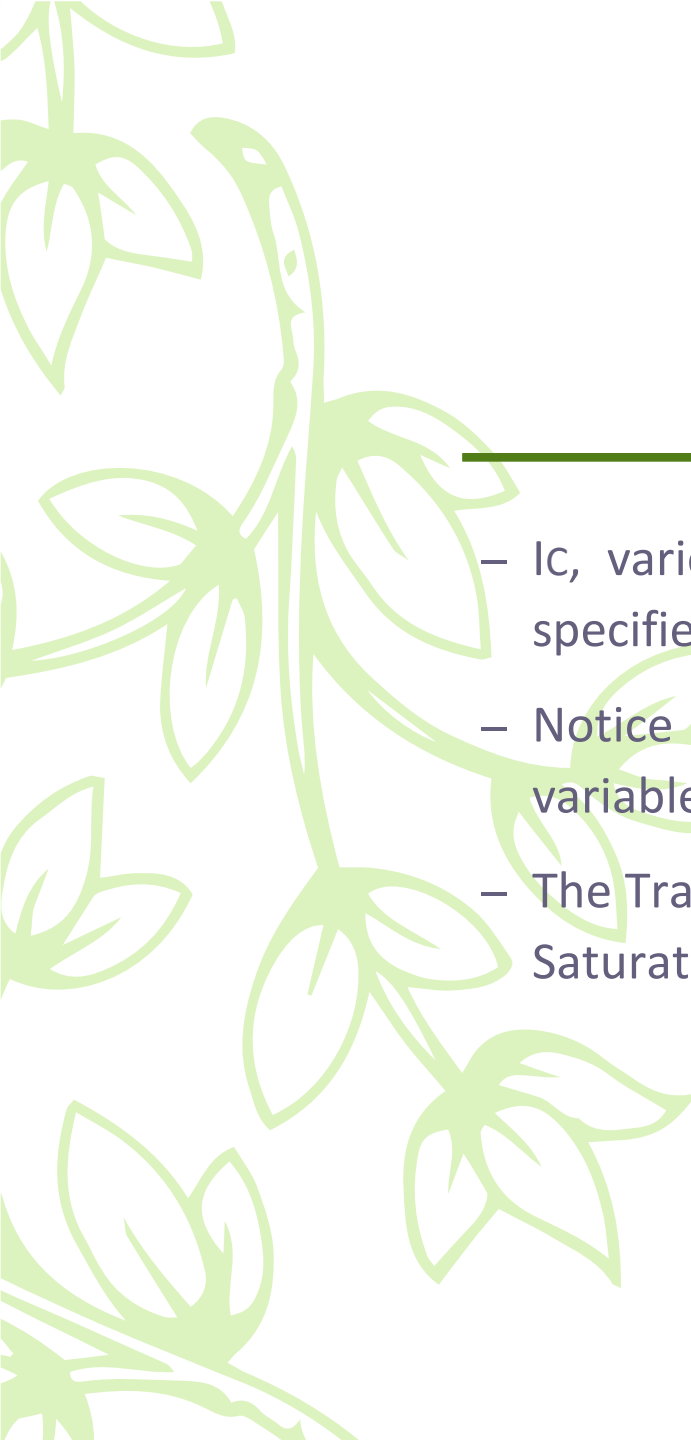
# Collector Characteristic Curves



(a) Circuit



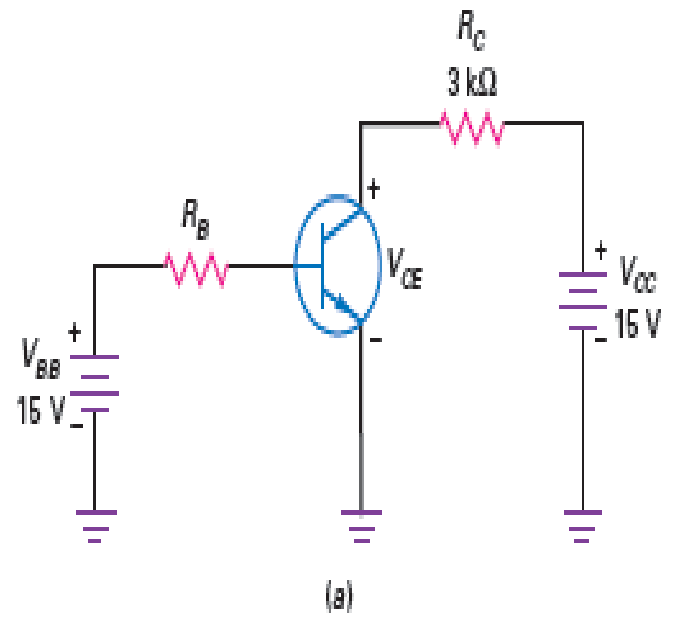
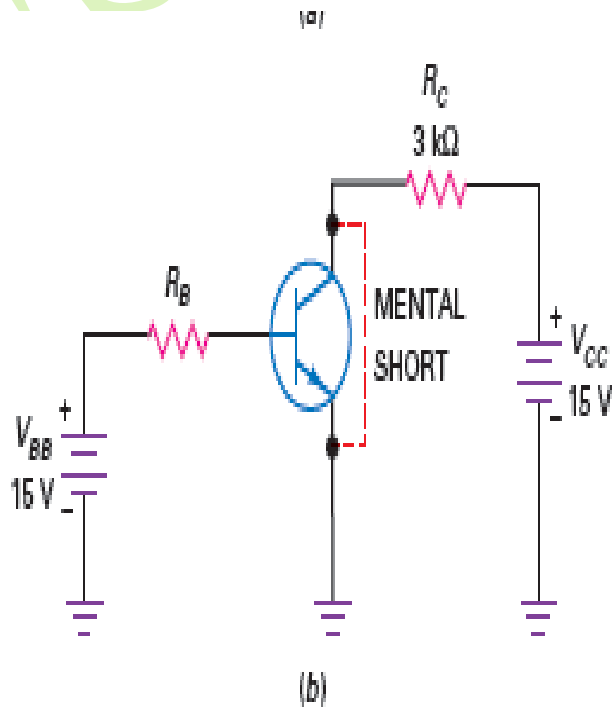
(b)  $I_C$  versus  $V_{CE}$  curve for one value of  $I_B$

- 
- 
- $I_C$ , varies with the collector-to emitter voltage,  $V_{CE}$ , for specified values of base current,  $I_B$ .
  - Notice in the circuit diagram that both  $V_{BB}$  and  $V_{CC}$  are variable sources of voltage.
  - The Transistor has three regions of operations: Cutoff, Saturation and Linear Region.

# Saturation:

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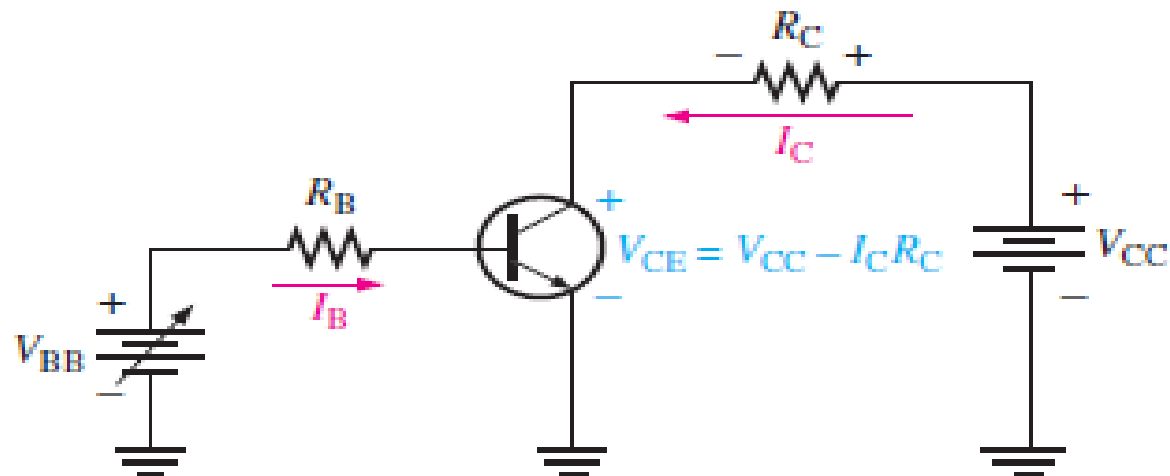
- When **both** junctions are forward-biased, the transistor is in the saturation region of its operation.
- **Saturation is the state of a BJT in which the collector current has reached a maximum and is independent of the base current.**
- When VCE reaches its saturation value,  $V_{CE(sat)}$ , the base-collector junction becomes forward-biased and  $I_C$  can increase no further even with a continued increase in  $I_B$ .
- $V_{CE(sat)}$  for a transistor occurs somewhere below the knee of the collector curves.



R.

- 
- **Saturation:** As  $I_B$  increases due to increasing  $V_{BB}$ ,  $I_C$  also increases and  $V_{CE}$  decreases due to the increased voltage drop across  $R_C$ . When the transistor reaches saturation,  $I_C$  can increase no further regardless of further increase in  $I_B$ . **Base-emitter and base-collector junctions are forward-biased.**

$$I_{C(sat)} = \frac{V_{CC}}{R_C}$$

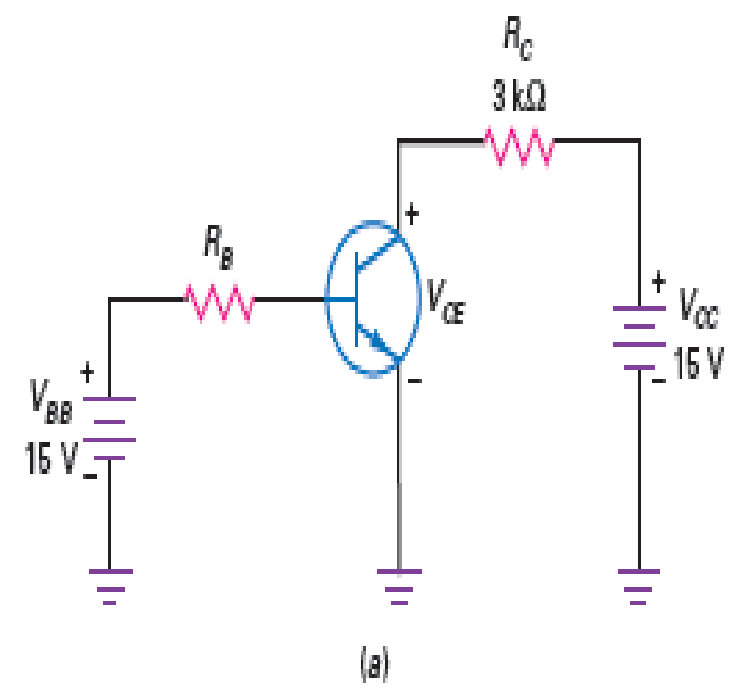
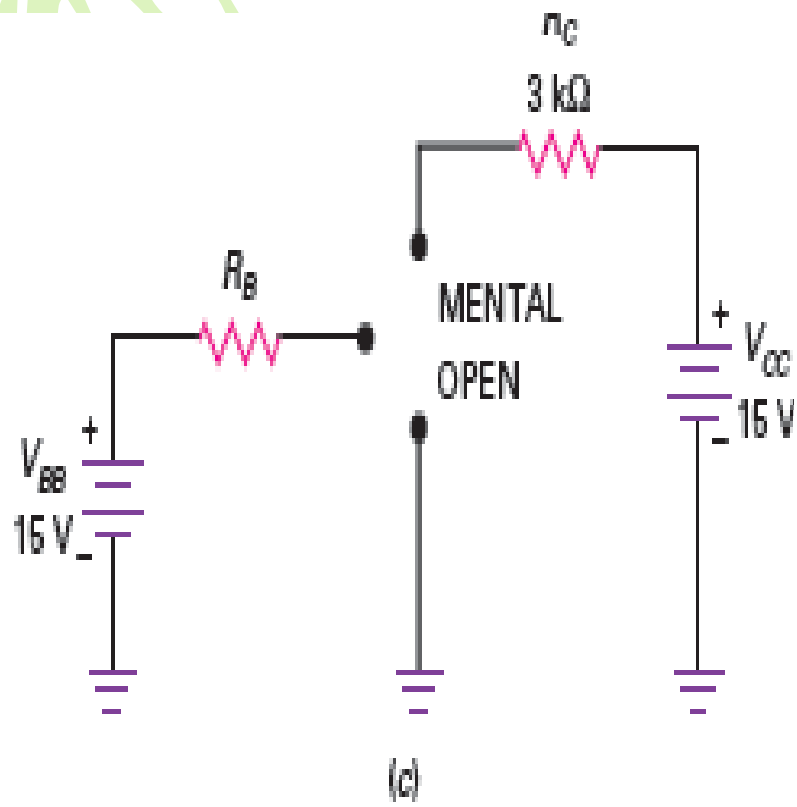


# Cutoff

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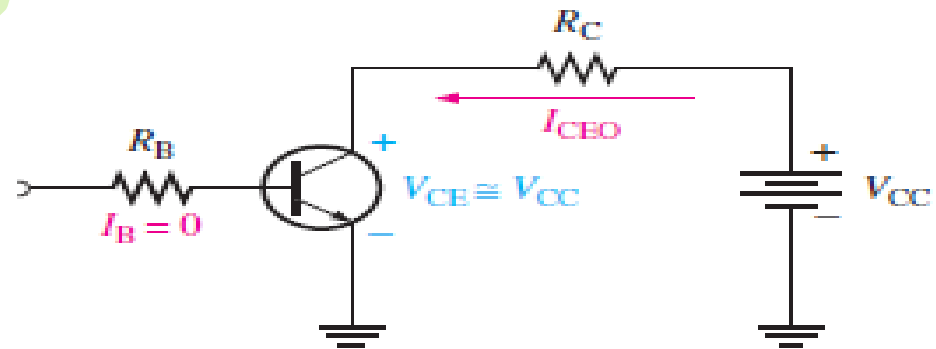
- when  $I_B = 0$ , the transistor is in the cutoff region of its operation.
- This is shown in the figure with the base lead open, resulting in a base current of zero. Under this condition, there is a very small amount of collector leakage current,  $I_{CEO}$ , due mainly to thermally produced carriers.
- Because  $I_{CEO}$  is extremely small, it will usually be neglected in circuit analysis so that  $V_{CE} = V_{CC}$ .



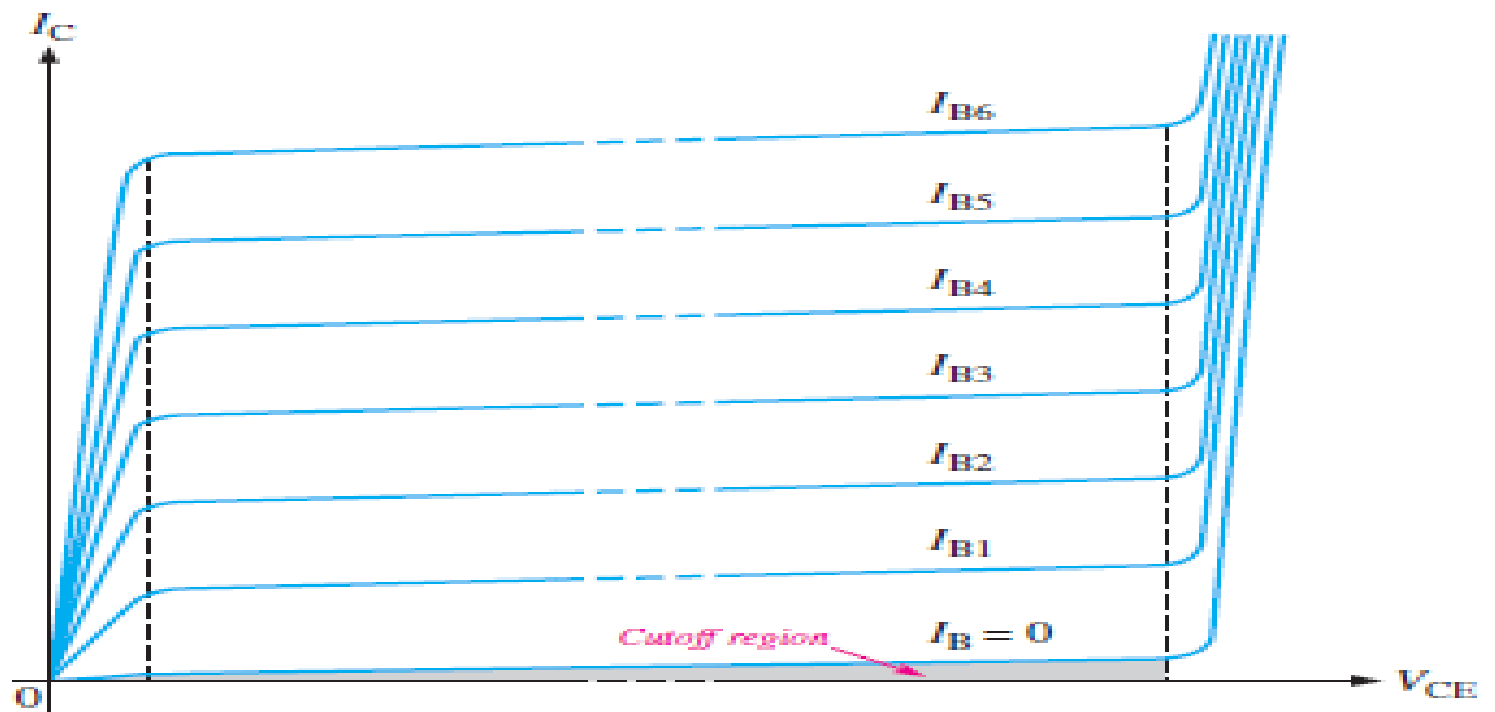


□

- 
- **In cutoff**, neither the base-emitter nor the base-collector junctions are forward-biased. The subscript CEO represents collector to-emitter with the **base open**.



**Cutoff:** Collector leakage current ( $I_{CEO}$ ) is extremely small and is usually neglected. **Base-emitter (BE) and base-collector junctions(BC) are reverse-biased.**

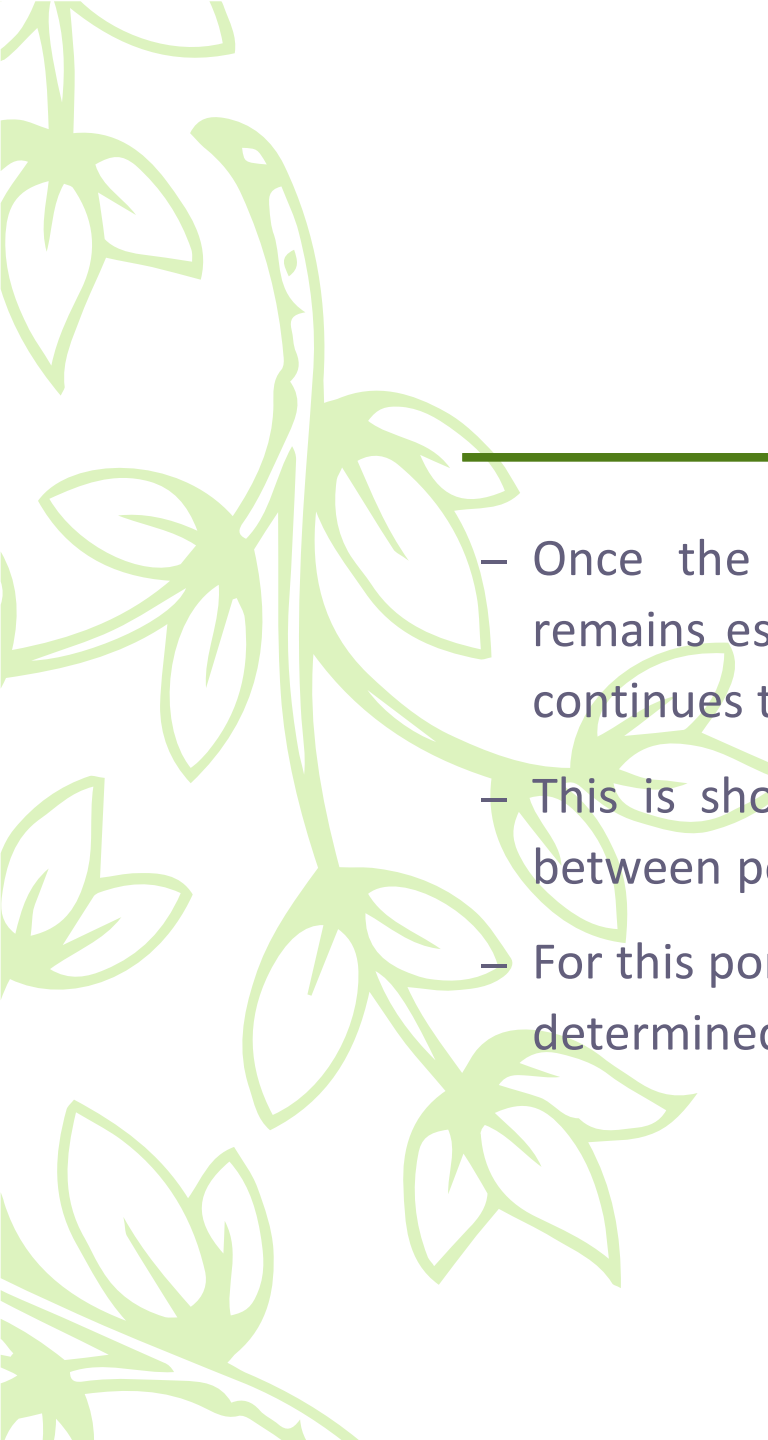


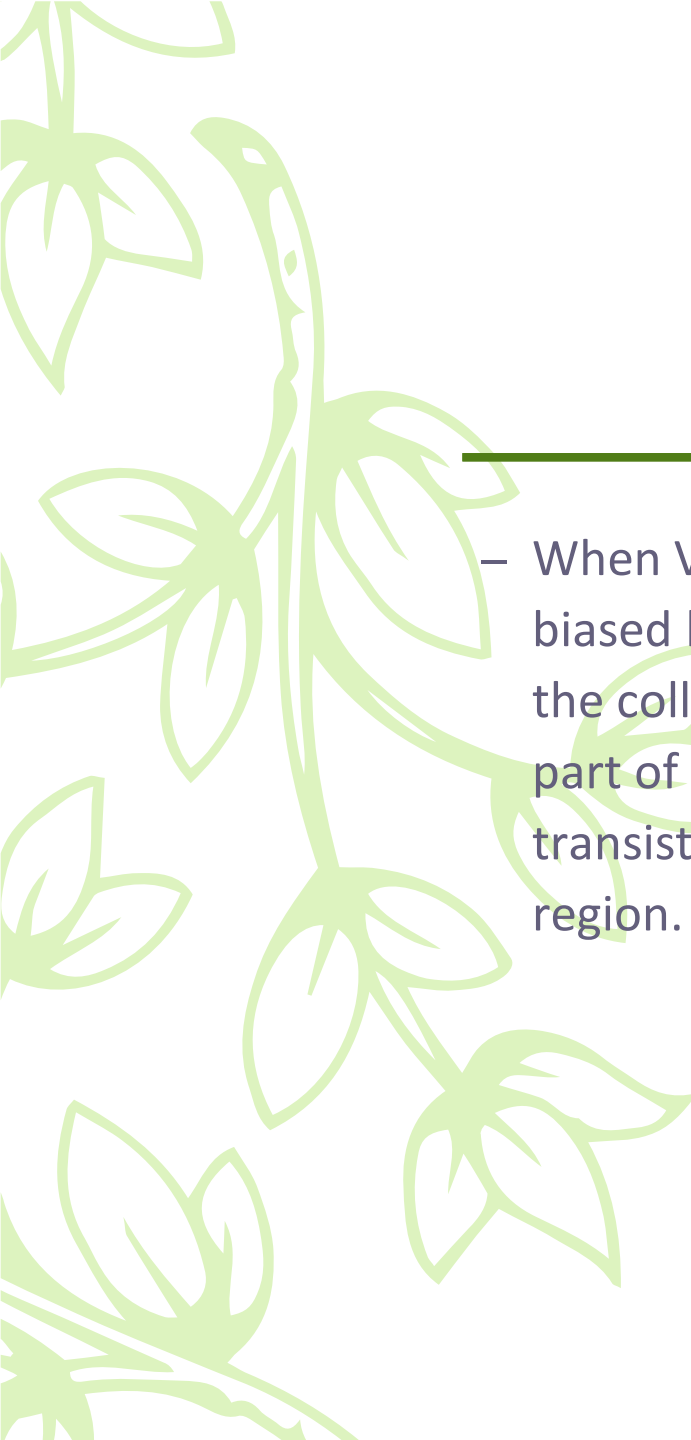
(c) Family of  $I_C$  versus  $V_{CE}$  curves for several values of  $I_B$  ( $I_{B1} < I_{B2} < I_{B3}$ , etc.)

# Linear or Active Region

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- As  $V_{CC}$  is increased,  $V_{CE}$  increases as the collector current increases. This is indicated by the portion of the characteristic curve between points A and B in Figure (b).
- $I_C$  increases as  $V_{CC}$  is increased because  $V_{CE}$  remains less than 0.7 V due to the forward-biased base-collector junction.
- Ideally, when  $V_{CE}$  exceeds 0.7 V, the base-collector junction becomes reverse-biased and the transistor goes into the active, or **linear, region of its operation**.

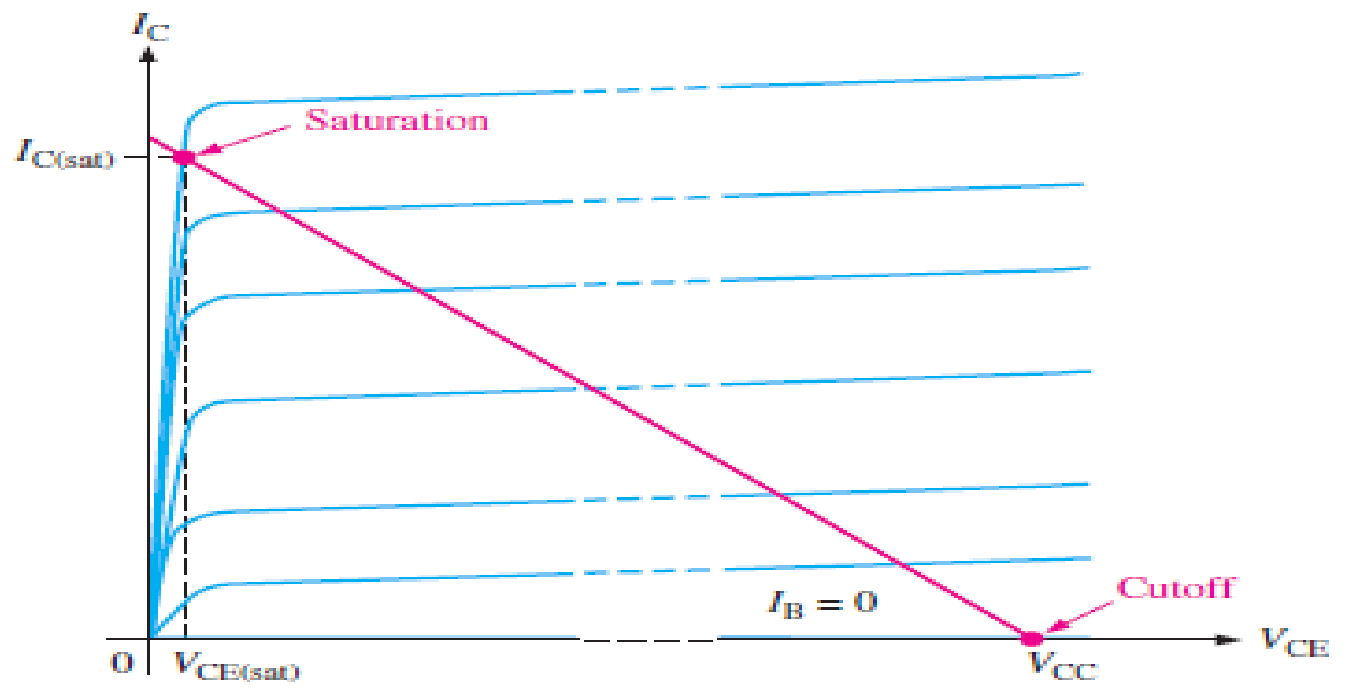
- 
- 
- Once the base collector junction is reverse-biased,  $I_c$  remains essentially constant for a given value of  $I_B$  as  $V_{CE}$  continues to increase.
  - This is shown by the portion of the characteristic curve between points B and C in Figure (b).
  - For this portion of the characteristic curve, the value of  $I_c$  is determined only by the relationship expressed as  $I_c = \beta_{DC} I_B$ .

- 
- 
- When  $V_{CE}$  reaches a sufficiently high voltage, the reverse-biased base-collector junction goes into breakdown; and the collector current increases rapidly as indicated by the part of the curve to the right of point C in Figure (b). A transistor should never be operated in this breakdown region.

# DC Load Line

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- Cutoff and saturation can be illustrated in relation to the collector characteristic curves by the use of a **load line**.
- **The bottom** of the load line is at ideal **cutoff** where  $I_C = 0$  and  $V_{CE} = V_{CC}$ . The top of the load line is at **saturation** where  $I_C = I_C(\text{sat})$  and  $V_{CE} = V_{CE}(\text{sat})$ .
- In between cutoff and saturation along the load line is the **active region** of the transistor's operation.





# Why is the load line useful?

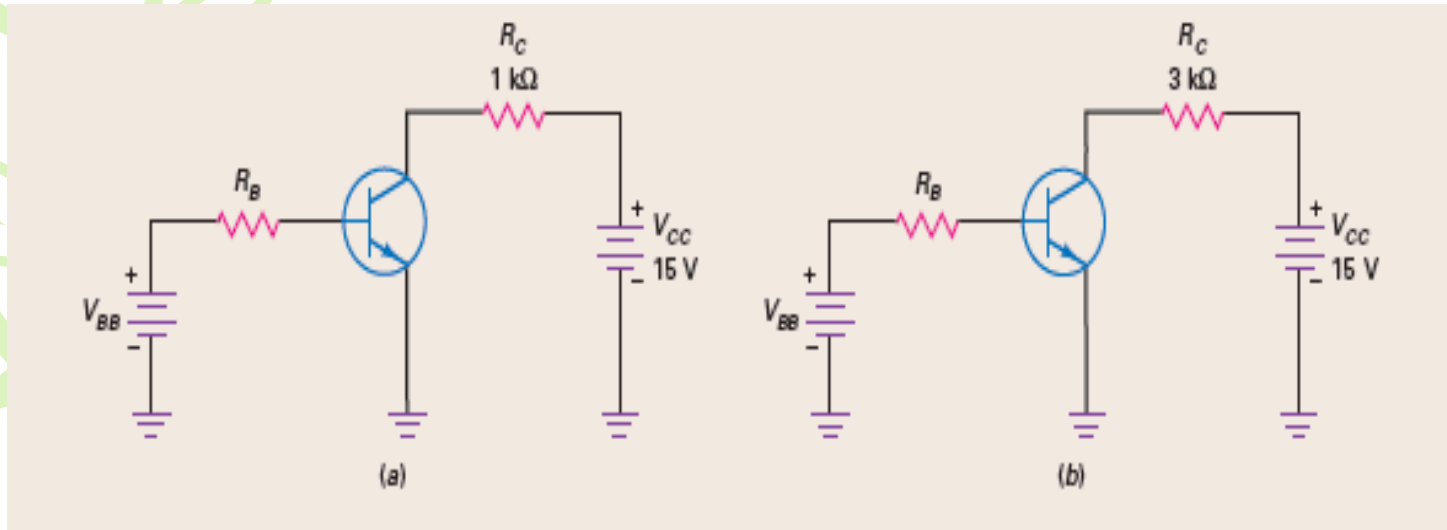
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Because it contains every possible operating point for the circuit. Stated another way, when the base resistance varies from zero to infinity, it causes  $I_B$  to vary, which makes  $I_C$  and  $V_{CE}$  vary over their entire ranges.

If you plot the  $I_C$  and  $V_{CE}$  values for every possible  $I_B$  value, you will get the load line. Therefore, the load line is a visual summary of *all possible transistor operating points*

# Example

- What are the saturation current and the cutoff voltage in Fig. *a* and Fig. *b*? Then, compare the load lines for both



# solution

Fig a

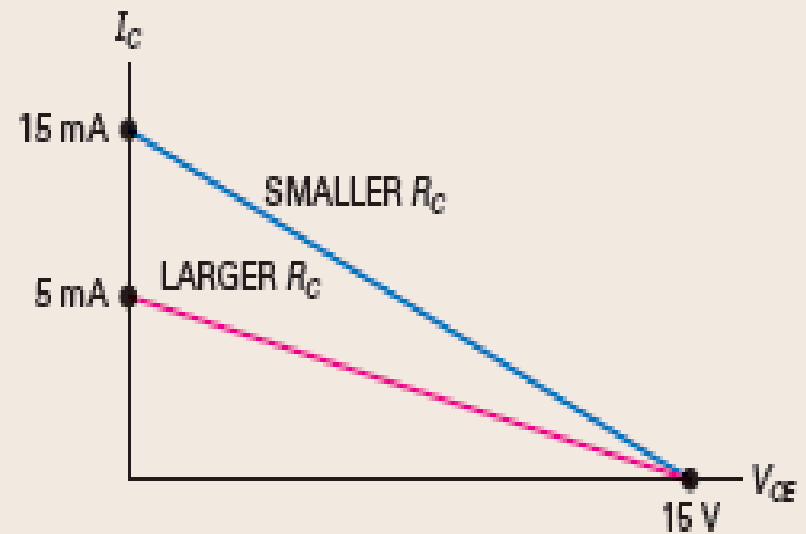
$$V_{CE(\text{cutoff})} = 15 \text{ V}$$

$$I_{C(\text{sat})} = \frac{15 \text{ V}}{1 \text{ k}\Omega} = 15 \text{ mA}$$

Fig b

$$V_{CE(\text{cutoff})} = 15 \text{ V}$$

$$I_{C(\text{sat})} = \frac{15 \text{ V}}{3 \text{ k}\Omega} = 5 \text{ mA}$$



(c)

# The Operating Point (Q-POINT)

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$$I_B = \frac{V_{BB} - V_{BE}}{R_B}$$

$$I_C = \beta_{dc} I_B$$

$$V_{CE} = V_{CC} - I_C R_C$$

### EXAMPLE 5-1

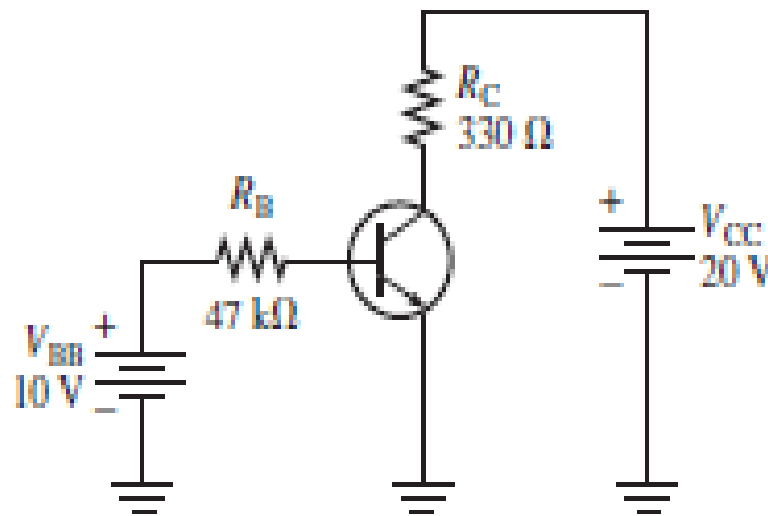
Determine the Q-point for the circuit in Figure 5-7 and draw the dc load line. Find the maximum peak value of base current for linear operation. Assume  $\beta_{DC} = 200$ .

**Solution** The Q-point is defined by the values of  $I_C$  and  $V_{CE}$ .

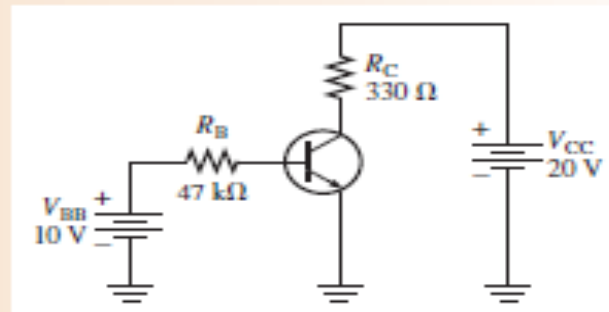
$$I_B = \frac{V_{BB} - V_{BE}}{R_B} = \frac{10 \text{ V} - 0.7 \text{ V}}{47 \text{ k}\Omega} = 198 \mu\text{A}$$

$$I_C = \beta_{DC} I_B = (200)(198 \mu\text{A}) = 39.6 \text{ mA}$$

$$V_{CE} = V_{CC} - I_C R_C = 20 \text{ V} - 13.07 \text{ V} = 6.93 \text{ V}$$



► FIGURE 5-7



The Q-point is at  $I_C = 39.6$  mA and at  $V_{CE} = 6.93$  V.

Since  $I_{C(\text{cutoff})} = 0$ , you need to know  $I_{C(\text{sat})}$  to determine how much variation in collector current can occur and still maintain linear operation of the transistor.

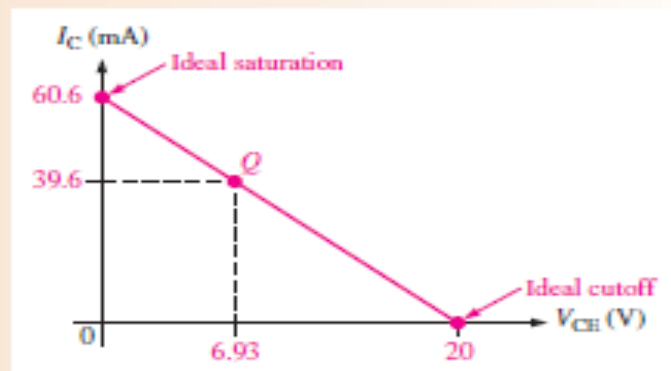
$$I_{C(\text{sat})} = \frac{V_{CC}}{R_C} = \frac{20 \text{ V}}{330 \Omega} = 60.6 \text{ mA}$$

The dc load line is graphically illustrated in Figure 5-8, showing that before saturation is reached,  $I_C$  can increase an amount ideally equal to

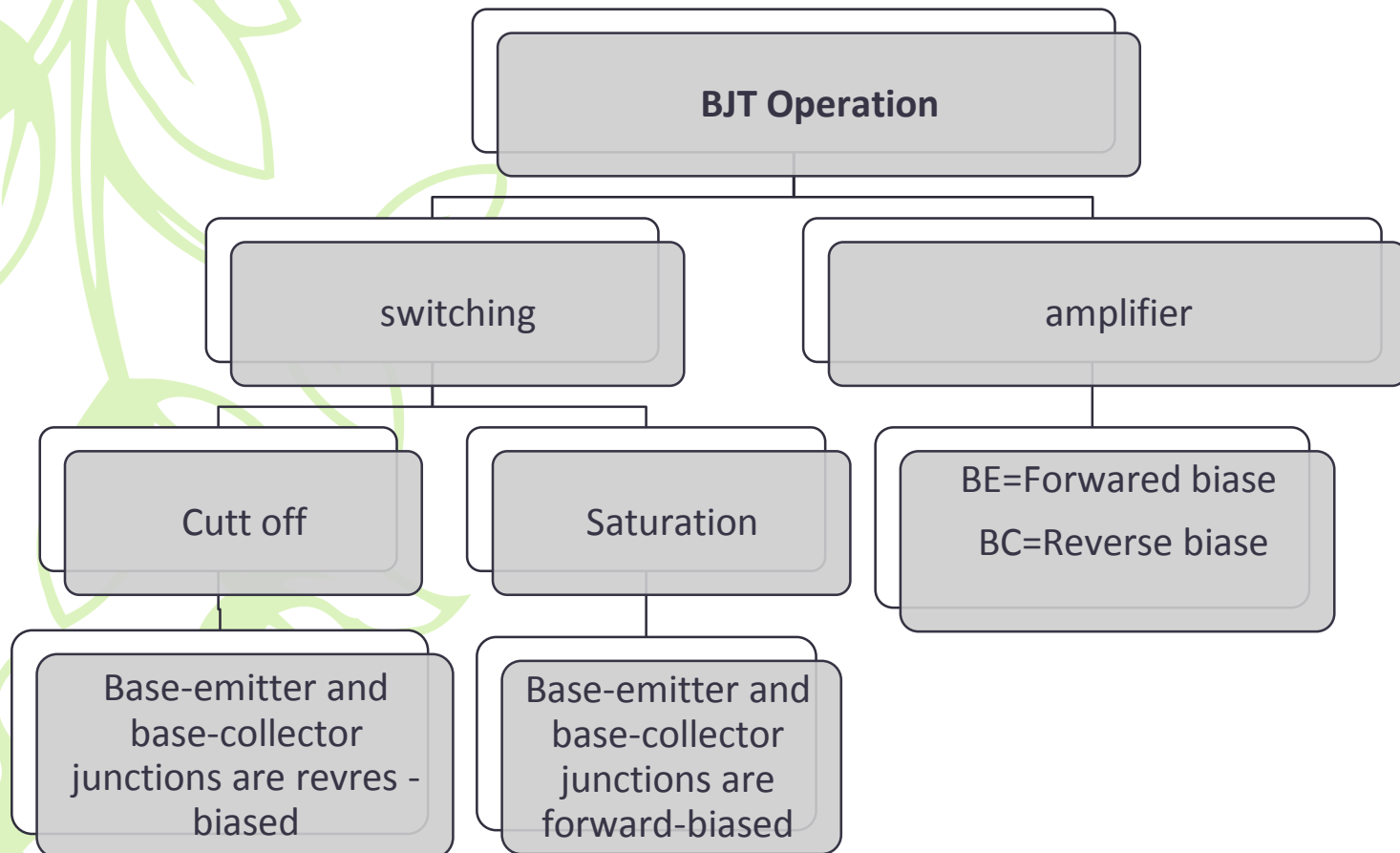
$$I_{C(\text{sat})} - I_{CQ} = 60.6 \text{ mA} - 39.6 \text{ mA} = 21.0 \text{ mA}$$

However,  $I_C$  can decrease by 39.6 mA before cutoff ( $I_C = 0$ ) is reached. Therefore, the limiting excursion is 21 mA because the Q-point is closer to saturation than to cutoff. The 21 mA is the maximum peak variation of the collector current. Actually, it would be slightly less in practice because  $V_{CE(\text{sat})}$  is not quite zero.

► FIGURE 5-8

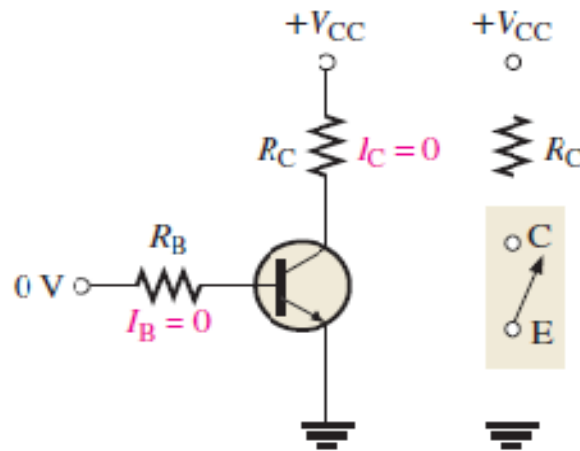


# BJT Operation

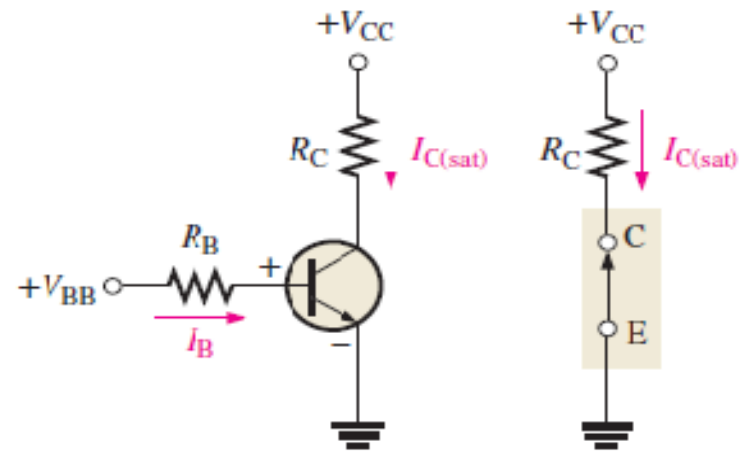


# The BJT as a switch

- The major application area is switching applications. When used as an electronic switch, a BJT is normally operated alternately in cutoff and saturation. Many digital circuits use the BJT as a switch



(a) Cutoff — open switch



(b) Saturation — closed switch



# Switching Operation

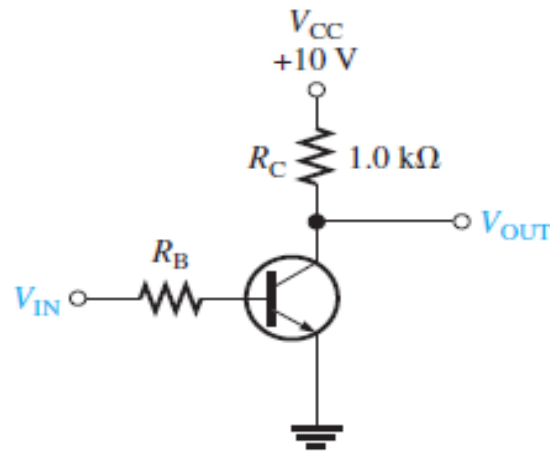
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- In part (a), the transistor is in the cutoff region because the base-emitter junction is not forward-biased. In this condition, there is, ideally, an open between collector and emitter.
- In part (b), the transistor is in the saturation region because the base emitter junction and the base-collector junction are forward-biased and the base current is made large enough to cause the collector current to reach its saturation value. In this condition, there is, ideally, a short between collector and emitter, as indicated by the switch equivalent.
- A small voltage drop across the transistor of up to a few tenths of a volt normally occurs, which is the saturation voltage,  $V_{CE(sat)}$ .

# Example

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- (a) For the transistor circuit in the figure, what is VCE when  $V_{IN} = 0$  V?
- (b) What minimum value of  $I_B$  is required to saturate this transistor if  $\beta_{DC}$  is 200? Neglect  $V_{CE(sat)}$ .
- (c) Calculate the maximum value of  $R_B$  when  $V_{IN} = 5$  V.



# Solution

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- (a) When  $V_{IN} = 0$  V, the transistor is in cutoff (acts like an open switch) and

$$V_{CE} = V_{CC} = 10 \text{ V}$$

- (b) Since  $V_{CE}(\text{sat})$  is neglected (assumed to be 0 V),

$$I_{C(\text{sat})} = V_{CC}/R_C = 10 \text{ V}/1.0 \text{ k}\Omega = 10 \text{ mA}$$

$$I_{B(\text{min})} = I_{C(\text{sat})}/\beta_{DC} = 10 \text{ mA}/200 = \mathbf{50 \mu A}$$

This is the value of  $I_B$  necessary to drive the to the point of saturation.

Any further increase in  $I_B$  will ensure the transistor remains in saturation but there cannot be any further increase in  $I_C$ .

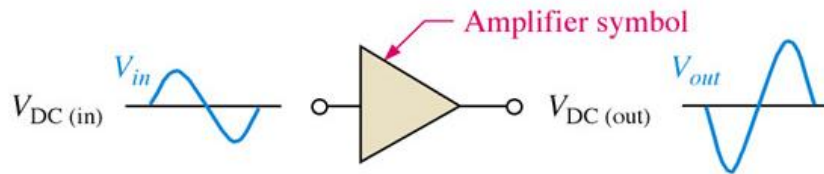
# Amplification

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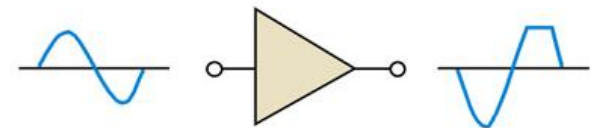
- The goal of amplification in most cases is to increase the amplitude of an ac signal without altering it.
- **Gain** The amount by which an electrical signal is increased or amplified
- Amplification is **the process of linearly increasing the amplitude of an electrical signal** and is one of the major properties of a transistor.
- When a BJT is biased in the active (or linear) region, as previously described, the BE junction has a low resistance due to forward bias and the BC junction has a high resistance due to reverse bias.

# BJT as an Amplifier

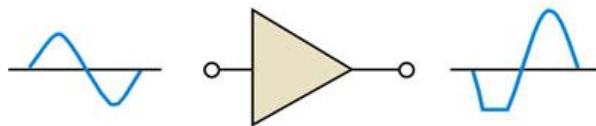
- A transistor amplifies current because the collector current is equal to the base current multiplied by the current gain  $\beta$ .



(a) Linear operation: larger output has same shape as input except that it is inverted

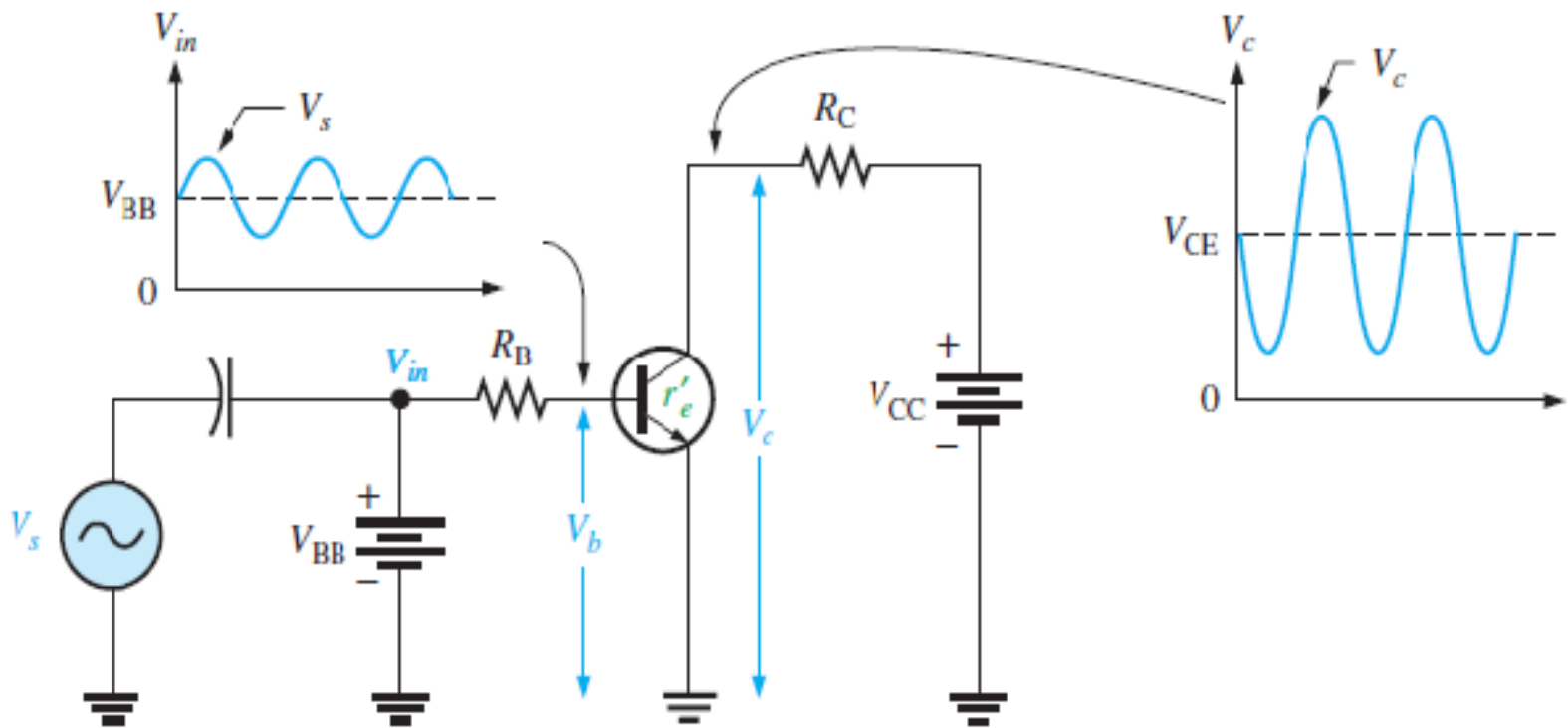


(b) Nonlinear operation: output voltage limited (clipped) by cutoff



(c) Nonlinear operation: output voltage limited (clipped) by saturation

# Voltage Amplifier





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**Thank you**